

基于事件触发的时滞 Lur'e 系统主从同步研究*

马健武, 高 燕, 顾凤蛟, 陈玲琦

(上海工程技术大学 电子电气工程学院, 上海 201600)

摘要: 研究了基于事件触发采样控制的时滞混沌 Lur'e 系统主从同步问题. 首先考虑了系统中包含的传输时滞构造了系统时滞模型. 然后, 通过构造三重积分项的 Lyapunov-Krasovskii 泛函, 并结合 Wirtinger 积分不等式和凸组合技术对 Lyapunov-Krasovskii 泛函的导数进行估计, 给出了混沌系统主从同步的充分条件. 所提出的事件触发机制应用于主从同步研究中, 可以有效地减少采样数据传输, 缓解网络带宽压力, 提高网络带宽利用率. 最后, 通过时滞蔡氏电路的数值仿真, 验证了所提出同步准则的有效性.

关键词: 混沌 Lur'e 系统; 主从同步; 事件触发; 采样数据控制

中图分类号: O175 文献标志码: A DOI: 10.21656/1000-0887.410228

Research on Master-Slave Synchronization of Time-Delay Lur'e Systems Based on Event Triggering

MA Jianwu, GAO Yan, GU Fengjiao, CHEN Lingqi

(School of Electronic and Electrical Engineering, Shanghai University of Engineering Science, Shanghai 201600, P.R.China)

Abstract: The master-slave synchronization of chaotic Lur'e systems with time-varying delays was investigated based on the event-triggered sampled-data control. First, the transmission time delays contained in the systems were considered, and the system time delay model was constructed. Then, the Lyapunov-Krasovskii functionals with triple integrals were constructed, the Wirtinger integral inequality and the convex combination technique were combined to estimate the derivatives of the Lyapunov-Krasovskii functionals, and the sufficient conditions for the master-slave synchronization of the chaotic system were given. Application of the proposed event-triggered mechanism to master-slave synchronization can effectively reduce the sampled data transmission, alleviate the pressure on the network bandwidth, and improve the utilization rate of the network bandwidth. Finally, the numerical simulation of the time-delay Chua's circuit verifies the effectiveness of the proposed synchronization criterion.

Key words: chaotic Lur'e system; master-slave synchronization; event-triggered; sampled-data control

引 言

时滞广泛存在于实际工程应用中, 且是不可避免的. 时滞的存在会对系统的动力学行为产生重要的影

* 收稿日期: 2020-08-03; 修订日期: 2020-10-27

作者简介: 马健武(1995—), 男, 硕士生(E-mail: 1596673894@qq.com);

高燕(1985—), 女, 讲师, 博士, 硕士生导师(通讯作者. E-mail: gy@sues.edu.cn).

引用格式: 马健武, 高燕, 顾凤蛟, 陈玲琦. 基于事件触发的时滞 Lur'e 系统主从同步研究[J]. 应用数学和力学, 2021, 42(7): 751-763.

响,往往会降低系统控制的性能,甚至破坏其稳定性,造成系统的不稳定.时滞系统动力学在航空航天、神经网络、通讯保密、生物生态学等领域得到了广泛的应用,因此,对时滞系统的研究具有理论意义和实际价值.

Lur'e 系统是一类非常典型的区间非线性系统,Chua's 电路、混沌神经网络、 n - 涡旋吸引子等非线性系统都可以用 Lur'e 系统进行建模.混沌 Lur'e 系统的同步研究引起了广大学者的兴趣^[1-4].已经有许多方法被提出用于实现混沌 Lur'e 系统同步,比如时滞反馈控制^[5-7]、自适应控制^[8-9]、脉冲控制^[10]、比例微分控制^[11-12]、采样数据控制^[13-16].在这些控制方法中,采样数据控制是数字控制,与其他模拟控制相比,它只利用系统在采样瞬间的采样信息给控制器,减少了同步数据的传输,提高了带宽的使用率.

混沌系统的同步研究虽然得到了很大的发展,仍然有改进的空间,比如混沌系统同步的研究结果大部分都是基于时间触发的,时间触发控制虽然易于实现,但可能会造成大量冗余的采样数据传输到带宽有限的通信网络中,造成网络堵塞.为了解决这些问题,事件触发的研究应运而生.对于事件触发机制而言,只有在系统状态满足触发条件时,事件触发机制才会向通信网络中传输数据.因此,引入事件触发机制可以显著减少采样数据传输、提高宽带利用率、节省通信资源.与时间触发控制相比,近年来事件触发控制更受学者关注.周期事件触发控制^[17-18]、自触发控制^[19]、自适应事件触发控制^[20-21]等事件触发机制相继被提出,在网络控制系统中备受青睐.

本文研究时滞 Lur'e 系统基于事件触发控制的同步问题,当满足触发条件时,释放当前采样数据到通信信道中并传输给控制器,否则将保持控制器中的数据不变,直至满足触发条件的采样数据到来.本文的创新点如下:

- 1) 本文在构造 Lyapunov-Krasovskii 泛函的过程中,考虑了时变时滞 $\tau(t)$ 的信息,引入了三重积分项和增广 Lyapunov-Krasovskii 泛函,考虑了 Lur'e 系统中非线性函数的可用信息.
- 2) 与传统的同步方法相比,本文在混沌 Lur'e 系统同步过程中加入了事件触发机制,能够有效地减少数据的传输,缓解通信带宽的压力,提高宽带利用率.
- 3) 本文在估计 Lyapunov-Krasovskii 泛函导数中出现的积分时引用了基于辅助函数的积分不等式,基于辅助函数的积分不等式包含了 Jensen 不等式和 Wirtinger 积分不等式,对交叉积分项的放大程度更接近于真实值,所以泛函导数也更接近于真实值,因此本文得到的同步判据保守性更低.

1 问题描述

考虑如下的主-从混沌 Lur'e 系统同步问题:

$$M: \begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{x}(t - \tau(t)) + \mathbf{E}f(\mathbf{D}\mathbf{x}(t)), \\ \mathbf{p}(t) = \mathbf{C}\mathbf{x}(t), \end{cases} \quad (1)$$

$$S: \begin{cases} \dot{\mathbf{y}}(t) = \mathbf{A}\mathbf{y}(t) + \mathbf{B}\mathbf{y}(t - \tau(t)) + \mathbf{E}f(\mathbf{D}\mathbf{y}(t)) + \mathbf{U}(t), \\ \mathbf{q}(t) = \mathbf{C}\mathbf{y}(t), \end{cases} \quad (2)$$

其中, M 是主系统, S 是从系统, $\mathbf{x}(t) \in \mathbf{R}^n$ 和 $\mathbf{y}(t) \in \mathbf{R}^n$ 是状态向量, $\mathbf{p}(t) \in \mathbf{R}^l$ 和 $\mathbf{q}(t) \in \mathbf{R}^l$ 是输出向量. $\mathbf{A} \in \mathbf{R}^{n \times n}$, $\mathbf{B} \in \mathbf{R}^{n \times n}$, $\mathbf{C} \in \mathbf{R}^{l \times n}$, $\mathbf{D} \in \mathbf{R}^{m \times n}$, $\mathbf{E} \in \mathbf{R}^{n \times m}$, 为已知常数矩阵, 时变时滞 $\tau(t)$ 满足

$$0 \leq \tau_1 \leq \tau(t) \leq \tau_2, \quad u_1 \leq \dot{\tau}(t) \leq u_2, \quad (3)$$

非线性函数 $f(\cdot) \triangleq [f_1(\cdot), f_2(\cdot), \dots, f_m(\cdot)] \in \mathbf{R}^m$, 满足 $f(0) = \mathbf{0}$, 假设 $f_i(\cdot)$ 属于扇形 $[k_i^-, k_i^+]$, 并满足条件:

$$[f_i(s) - k_i^+ s][f_i(s) - k_i^-] \leq 0, \quad \forall s \in \mathbf{R}, \quad i = 1, 2, \dots, m. \quad (4)$$

本文通过采样数据反馈研究了主从同步问题,系统的输出向量 $\mathbf{p}(t), \mathbf{q}(t)$ 的离散测量用于实现控制目的,在采样时刻 t_k 的离散测量为 $\mathbf{p}(t_k), \mathbf{q}(t_k)$ ($k \in \mathbf{N}$), 在传统的控制中,假设控制信号是由零阶保持器产生的,其保持时间序列为 $[t_k, t_{k+1})$. 在本文中, t_k 是从传感器到采样器的传输时刻,采样控制输入可以采用如下形式表示:

$$\mathbf{U}(t) = \mathbf{K}[\mathbf{p}(t_k) - \mathbf{q}(t_k)], \quad t \in [t_k, t_{k+1}). \quad (5)$$

误差系统定义为

$$\dot{\boldsymbol{e}}(t) = \dot{\boldsymbol{x}}(t) - \dot{\boldsymbol{y}}(t) = \boldsymbol{A}\boldsymbol{e}(t) + \boldsymbol{B}\boldsymbol{e}(t - \tau(t)) + \boldsymbol{E}\boldsymbol{\eta}(\boldsymbol{D}\boldsymbol{e}(t)) - \boldsymbol{U}(t). \quad (6)$$

下面分两种情况分析网络传输的时滞.

情况 1 如果 $t_k + h_1 + \bar{\tau} \geq t_{k+1} + \tau_{k+1}$, 定义

$$h(t) = t - t_k, \quad t \in [t_k + \tau_k, t_{k+1} + \tau_{k+1}), \quad (7)$$

显然 $\tau_k \leq h(t) \leq t_{k+1} - t_k + \tau_{k+1} \leq h_1 + \bar{\tau}$.

情况 2 如果 $t_k + h_1 + \bar{\tau} < t_{k+1} + \tau_{k+1}$, 考虑把 $[t_k + \tau_k, t_{k+1} + \tau_{k+1})$ 进行如下分段:

$$\begin{cases} I_1 = [t_k + \tau_k, t_k + h_1 + \bar{\tau}), \\ I_j = \left[t_k + \sum_{i=1}^{j-1} h_i + \bar{\tau}, t_k + \sum_{i=1}^j h_i + \bar{\tau} \right), \quad j = 2, 3, \dots, M-1, \\ I_{M-1} = \left[t_k + \sum_{i=1}^{M-1} h_i + \bar{\tau}, t_{k+1} + \tau_{k+1} \right). \end{cases} \quad (8)$$

分区网络延迟函数 $h(t)$ 定义如下:

$$h(t) = \begin{cases} t - t_k, & t \in I_1, \\ t - t_k - \sum_{i=1}^{j-1} h_i, & t \in I_j, j = 2, 3, \dots, M-2, \\ t - t_k - \sum_{i=1}^{M-1} h_i, & t \in I_{M-1}. \end{cases} \quad (9)$$

由此可得, $0 \leq \tau_k \leq h(t) \leq \bar{h} + \bar{\tau} \triangleq \bar{h}$, $t \in [t_k + \tau_k, t_{k+1} + \tau_{k+1})$, 其中, $\bar{h} = \max\{h_1, h_2, \dots, h_M\}$.

当前采样数据 $\boldsymbol{e}(t_s)$ 和上一次传输数据 $\boldsymbol{e}(t_k)$ 之差 $\boldsymbol{\varepsilon}(t_s)$ 定义如下:

$$\boldsymbol{\varepsilon}(t_s) = \begin{cases} \mathbf{0}, & t \in I_1, \\ \boldsymbol{e}\left(t_k + \sum_{i=1}^{j-1} h_i\right) - \boldsymbol{e}(t_k), & t \in I_j, j = 2, 3, \dots, M-2, \\ \boldsymbol{e}\left(t_k + \sum_{i=1}^{M-1} h_i\right) - \boldsymbol{e}(t_k), & t \in I_{M-1}. \end{cases} \quad (10)$$

当前采样数据是否传输给控制器取决于事件触发条件: 当前采样数据满足触发条件时, 将会被传输到控制器; 当前采样数据不满足触发条件时, 将保持传感器的数据不变. 本文所设计的事件触发条件如下:

$$\boldsymbol{\varepsilon}^T(t_s) \boldsymbol{\Theta}_1 \boldsymbol{\varepsilon}(t_s) - \sigma_1(t) \boldsymbol{e}^T(t_s) \boldsymbol{\Theta}_2 \boldsymbol{e}(t_s) - \sigma_2(t) \boldsymbol{e}^T(t_k) \boldsymbol{\Theta}_3 \boldsymbol{e}(t_k) \leq 0, \quad (11)$$

其中, $t_s = t_k + \sum_{i=1}^n h_i$ 是当前采样时刻, t_k 是最近传输时刻, $\boldsymbol{\varepsilon}(t_s) = \boldsymbol{e}(t_s) - \boldsymbol{e}(t_k)$, 矩阵 $\boldsymbol{\Theta}_i (i = 1, 2, 3)$ 是正定对称矩阵, 函数 $\sigma_1(t), \sigma_2(t)$ 满足

$$\dot{\sigma}_1(t) = \frac{1}{\sigma_1(t)} \left[\frac{1}{\sigma_1(t)} - \mu_1 \right] [\boldsymbol{\varepsilon}^T(t_s) \boldsymbol{\Theta}_1 \boldsymbol{\varepsilon}(t_s) - \kappa_1 \boldsymbol{e}^T(t_k) \boldsymbol{\Theta}_3 \boldsymbol{e}(t_k)], \quad (12)$$

$$\dot{\sigma}_2(t) = \frac{1}{\sigma_2(t)} \left[\frac{1}{\sigma_2(t)} - \mu_2 \right] [\boldsymbol{\varepsilon}^T(t_s) \boldsymbol{\Theta}_1 \boldsymbol{\varepsilon}(t_s) - \kappa_2 \boldsymbol{e}^T(t_s) \boldsymbol{\Theta}_2 \boldsymbol{e}(t_s)], \quad (13)$$

式中, $\mu_1, \mu_2, \kappa_1, \kappa_2$ 是给定的正常数, 使得 $1/\mu_1 < \sigma_1(t) < \kappa_2, 1/\mu_2 < \sigma_2(t) < \kappa_1$.

为了获得混沌 Lur'e 系统主从同步的判据, 需要引入如下引理.

引理 1^[22] 对于正定矩阵 $\boldsymbol{R} > \mathbf{0}$, 和可微函数 $\{\boldsymbol{\omega}(s) \mid s \in [a, b]\}$, 有如下不等式成立:

$$\int_a^b \int_{\theta}^b \dot{\boldsymbol{\omega}}^T(s) \boldsymbol{R} \dot{\boldsymbol{\omega}}(s) ds d\theta \geq 2\boldsymbol{\Omega}_3^T \boldsymbol{R} \boldsymbol{\Omega}_3 + 4\boldsymbol{\Omega}_4^T \boldsymbol{R} \boldsymbol{\Omega}_4,$$

$$\int_a^b \int_a^{\theta} \dot{\boldsymbol{\omega}}^T(s) \boldsymbol{R} \dot{\boldsymbol{\omega}}(s) ds d\theta \geq 2\boldsymbol{\Omega}_5^T \boldsymbol{R} \boldsymbol{\Omega}_5 + 4\boldsymbol{\Omega}_6^T \boldsymbol{R} \boldsymbol{\Omega}_6,$$

其中

$$\boldsymbol{\Omega}_3 = \boldsymbol{\omega}(b) - \frac{1}{b-a} \int_a^b \boldsymbol{\omega}(s) ds, \quad \boldsymbol{\Omega}_4 = \boldsymbol{\omega}(b) + \frac{2}{b-a} \int_a^b \boldsymbol{\omega}(s) ds - \frac{6}{(b-a)^2} \int_a^b \int_{\theta}^b \boldsymbol{\omega}(s) ds d\theta,$$

$$\Omega_5 = \omega(b) - \frac{1}{b-a} \int_a^b \omega(s) ds, \quad \Omega_6 = \omega(b) - \frac{4}{b-a} \int_a^b \omega(s) ds + \frac{6}{(b-a)^2} \int_a^b \int_\theta^b \omega(s) ds d\theta.$$

引理 2^[23] 给定一个正定矩阵 $R > \mathbf{0}$, 对于连续函数 $\{\omega(s) \mid s \in [a, b]\}$, 有如下不等式成立:

$$\int_a^b \omega^T(u) R \omega(u) du \geq \frac{1}{b-a} \Omega_7^T R \Omega_7 + \frac{3}{b-a} \Omega_8^T R \Omega_8,$$

其中

$$\Omega_7 = \int_a^b \omega(u) du, \quad \Omega_8 = \int_a^b \omega(u) du - \frac{2}{b-a} \int_a^b \int_a^s \omega(r) dr ds.$$

引理 3^[24] 对于向量 ω , 标量 $a \leq b$, 对称矩阵 $R > \mathbf{0}$, 有如下不等式成立:

$$\int_a^b \dot{\omega}^T(s) R \dot{\omega}(s) ds \geq \frac{1}{b-a} \chi_1^T R \chi_1 + \frac{3}{b-a} \chi_2^T R \chi_2 + \frac{5}{b-a} \chi_3^T R \chi_3,$$

其中

$$\chi_1 = \omega(b) - \omega(a), \quad \chi_2 = \omega(b) + \omega(a) - \frac{2}{b-a} \int_a^b \omega(s) ds,$$

$$\chi_3 = \chi_1 + \frac{6}{b-a} \int_a^b \omega(s) ds - \frac{12}{(b-a)^2} \int_a^b \int_\theta^b \omega(s) ds d\theta.$$

引理 4^[25] 对于一个实标量 $\alpha \in (0, 1)$, 对称矩阵 $X_1 > \mathbf{0}$ 和 $X_2 > \mathbf{0}$, 任意矩阵 S_1 和 S_2 , 使得以下矩阵不等式成立:

$$\begin{pmatrix} \frac{1}{\alpha} X_1 & \mathbf{0} \\ \mathbf{0} & \frac{1}{1-\alpha} X_2 \end{pmatrix} \geq \begin{pmatrix} X_1 + (1-\alpha) P_1 & (1-\alpha) S_1 + \alpha S_2 \\ * & X_2 + \alpha P_2 \end{pmatrix},$$

其中 $P_1 = X_1 - S_2 X_2^{-1} S_2^T$, $P_2 = X_2 - S_1^T X_1^{-1} S_1$.

为了简化文章的证明过程, 首先给出相应的记号:

$$\xi(t) = \text{col} \left\{ \mathbf{e}(t), \mathbf{e}(t - \tau_1), \mathbf{e}(t - \tau(t)), \mathbf{e}(t - \tau_2), \mathbf{e}(t - d), \mathbf{e}(t - h(t)), \mathbf{e}(t - h), \boldsymbol{\varepsilon}(t_s), \right.$$

$$\dot{\mathbf{e}}(t), \mathbf{g}(D\mathbf{e}(t)), \frac{1}{\tau_1} \int_{t-\tau_1}^t \mathbf{e}(s) ds, \frac{1}{\tau(t) - \tau_1} \int_{t-\tau(t)}^{t-\tau_1} \mathbf{e}(s) ds, \frac{1}{\tau_2 - \tau(t)} \int_{t-\tau_2}^{t-\tau(t)} \mathbf{e}(s) ds,$$

$$\frac{1}{d} \int_{t-d}^t \mathbf{e}(s) ds, \frac{1}{h(t) - d} \int_{t-h(t)}^{t-d} \mathbf{e}(s) ds, \frac{1}{h - h(t)} \int_{t-h}^{t-h(t)} \mathbf{e}(s) ds, \frac{2}{\tau_1^2} \int_{t-\tau_1}^t \int_s^t \mathbf{e}(\theta) d\theta ds,$$

$$\frac{2}{[\tau(t) - \tau_1]^2} \int_{t-\tau(t)}^{t-\tau_1} \int_s^{t-\tau_1} \mathbf{e}(\theta) d\theta ds, \frac{2}{[\tau_2 - \tau(t)]^2} \int_{t-\tau_2}^{t-\tau(t)} \int_s^{t-\tau(t)} \mathbf{e}(\theta) d\theta ds, \frac{2}{d^2} \int_{t-d}^t \int_s^t \mathbf{e}(\theta) d\theta ds,$$

$$\frac{2}{[h(t) - d]^2} \int_{t-h(t)}^{t-d} \int_s^{t-d} \mathbf{e}(\theta) d\theta ds, \frac{2}{[h - h(t)]^2} \int_{t-h}^{t-h(t)} \int_s^{t-h(t)} \mathbf{e}(\theta) d\theta ds, \dot{\mathbf{e}}(t - \tau_1),$$

$$\dot{\mathbf{e}}(t - \tau(t)), \dot{\mathbf{e}}(t - \tau_2), \dot{\mathbf{e}}(t - d), \dot{\mathbf{e}}(t - h) \left. \right\},$$

$$\eta(t) = \text{col} \left\{ \mathbf{e}(t), \mathbf{e}(t - \tau_1), \mathbf{e}(t - \tau_2), \int_{t-\tau_2}^t \mathbf{e}(s) ds, \mathbf{e}(t - d), \mathbf{e}(t - h), \int_{t-h}^t \mathbf{e}(s) ds \right\},$$

$$\dot{\eta}(t) = \text{col} \left\{ \dot{\mathbf{e}}(t), \dot{\mathbf{e}}(t - \tau_1), \dot{\mathbf{e}}(t - \tau_2), \dot{\mathbf{e}}(t) - \dot{\mathbf{e}}(t - \tau_2), \dot{\mathbf{e}}(t - d), \dot{\mathbf{e}}(t - h), \dot{\mathbf{e}}(t) - \dot{\mathbf{e}}(t - h) \right\},$$

$$\chi(t) = \frac{1}{\tau_1} \int_{t-\tau_1}^t \mathbf{e}(s) ds, \quad \delta(t) = \frac{1}{\tau(t) - \tau_1} \int_{t-\tau(t)}^{t-\tau_1} \mathbf{e}(s) ds, \quad \boldsymbol{\varepsilon}(t) = \frac{1}{\tau_2 - \tau(t)} \int_{t-\tau_2}^{t-\tau(t)} \mathbf{x}(s) ds,$$

$$\phi(t) = \frac{1}{d} \int_{t-d}^t \mathbf{e}(s) ds, \quad \varphi(t) = \frac{1}{h(t) - d} \int_{t-h(t)}^{t-d} \mathbf{e}(s) ds, \quad \gamma(t) = \frac{1}{h - h(t)} \int_{t-h}^{t-h(t)} \mathbf{e}(s) ds,$$

$$\kappa(t) = \frac{2}{\tau_1^2} \int_{t-\tau_1}^t \int_s^t \mathbf{e}(\theta) d\theta ds, \quad \lambda(t) = \frac{2}{[\tau(t) - \tau_1]^2} \int_{t-\tau(t)}^{t-\tau_1} \int_s^{t-\tau_1} \mathbf{x}(\theta) d\theta ds,$$

$$\begin{aligned} \boldsymbol{\nu}(t) &= \frac{2}{[\tau_2 - \tau(t)]^2} \int_{t-\tau_2}^{t-\tau(t)} \int_s^{t-\tau(t)} \mathbf{e}(\theta) d\theta ds, \boldsymbol{\omega}(t) = \frac{2}{d^2} \int_{t-d}^t \int_s^t \mathbf{e}(\theta) d\theta ds, \\ \boldsymbol{\theta}(t) &= \frac{2}{[h(t) - d]^2} \int_{t-h(t)}^{t-d} \int_s^{t-d} \mathbf{x}(\theta) d\theta ds, \boldsymbol{\vartheta}(t) = \frac{2}{[h - h(t)]^2} \int_{t-h}^{t-h(t)} \int_s^{t-h(t)} \mathbf{e}(\theta) d\theta ds, \\ \mathbf{e}_i &= (\mathbf{0}_{n \times (i-1)n} \quad \mathbf{I}_n \quad \mathbf{0}_{n \times [(26-i)n+m]}), \quad i = 1, 2, \dots, 9, \\ \mathbf{e}_{10} &= (\mathbf{0}_{m \times 9n} \quad \mathbf{I}_m \quad \mathbf{0}_{m \times 17n}), \mathbf{e}_i = (\mathbf{0}_{n \times [(i-2)n+m]} \quad \mathbf{I}_n \quad \mathbf{0}_{n \times (27-i)n}), \quad i = 11, 12, \dots, 27, \\ \mathbf{L} &= \text{col} \{ \mathbf{e}_9 \quad \mathbf{e}_{23} \quad \mathbf{e}_{25} \quad \mathbf{e}_1 - \mathbf{e}_4 \quad \mathbf{e}_{26} \quad \mathbf{e}_{27} \quad \mathbf{e}_1 - \mathbf{e}_7 \}, \mathbf{L}_1 = (\mathbf{e}_1^T \quad \mathbf{e}_2^T \quad \mathbf{e}_4^T \quad \tau_1 \mathbf{e}_{11}^T \quad \mathbf{e}_5^T \quad \mathbf{e}_7^T \quad d\mathbf{e}_{14}^T), \\ \mathbf{L}_2 &= (\mathbf{0}_{26n+m, 6n} \quad \mathbf{e}_{12}^T \quad \mathbf{0}_{26n+m, 3n}), \mathbf{L}_3 = (\mathbf{0}_{26n+m, 3n} \quad \mathbf{e}_{13}^T \quad \mathbf{0}_{26n+m, 3n}), \\ \mathbf{L}_4 &= (\mathbf{0}_{26n+m, 6n} \quad \mathbf{e}_{15}^T), \mathbf{L}_5 = (\mathbf{0}_{26n+m, 6n} \quad \mathbf{e}_{16}^T), \\ \mathbf{H}_1 &= \begin{pmatrix} \mathbf{e}_2 - \mathbf{e}_3 \\ \mathbf{e}_2 + \mathbf{e}_3 - 2\mathbf{e}_{12} \\ \mathbf{e}_2 - \mathbf{e}_3 + 6\mathbf{e}_{12} - 6\mathbf{e}_{18} \end{pmatrix}, \mathbf{H}_2 = \begin{pmatrix} \mathbf{e}_3 - \mathbf{e}_4 \\ \mathbf{e}_3 + \mathbf{e}_4 - 2\mathbf{e}_{13} \\ \mathbf{e}_3 - \mathbf{e}_4 + 6\mathbf{e}_{13} - 6\mathbf{e}_{19} \end{pmatrix}, \\ \mathbf{H}_3 &= \begin{pmatrix} \mathbf{e}_5 - \mathbf{e}_6 \\ \mathbf{e}_5 + \mathbf{e}_6 - 2\mathbf{e}_{15} \\ \mathbf{e}_5 - \mathbf{e}_6 + 6\mathbf{e}_{15} - 6\mathbf{e}_{21} \end{pmatrix}, \mathbf{H}_4 = \begin{pmatrix} \mathbf{e}_6 - \mathbf{e}_7 \\ \mathbf{e}_6 + \mathbf{e}_7 - 2\mathbf{e}_{16} \\ \mathbf{e}_6 - \mathbf{e}_7 + 6\mathbf{e}_{16} - 6\mathbf{e}_{22} \end{pmatrix}. \end{aligned}$$

2 主要结果

本文在这一节中,给出使混沌 Lur'e 系统基于事件触发机制的采样控制器设计实现主从同步的方法,建立闭环系统(6)收敛的充分条件.

定理 1 对于给定的标量 $h > 0, \tau_2 \geq \tau_1 \geq 0, \mu_i > 0, \kappa_i > 0 (i = 1, 2), u_1, u_2$, 如果存在适当维数的矩阵 $\mathbf{P} > \mathbf{0}, \mathbf{Q}_i > \mathbf{0} (i = 1, 2, \dots, 6), \mathbf{R}_i > \mathbf{0} (i = 1, 2, \dots, 6), \mathbf{J}_i > \mathbf{0} (i = 1, 2, \dots, 8), \mathbf{S}_i > \mathbf{0} (i = 1, 2, 3, 4), \mathbf{T}_i > \mathbf{0} (i = 1, 2, 3, 4), \boldsymbol{\Theta}_i > \mathbf{0} (i = 1, 2, 3), \bar{\mathbf{T}}_i = \text{diag}(\mathbf{T}_i, 3\mathbf{T}_i, 5\mathbf{T}_i) (i = 1, 2, 3, 4), \bar{\mathbf{J}}_i = \text{diag}(\mathbf{J}_i, 3\mathbf{J}_i, 5\mathbf{J}_i) (i = 1, 2, 3, 4)$, 任意矩阵 $\mathbf{P}_i (i = 1, 2, 3, 4)$, 使得下列线性矩阵不等式是可行的:

$$\begin{pmatrix} \boldsymbol{\Psi}_1 + \boldsymbol{\Psi}_2 + \boldsymbol{\Psi}_{3,1} + \boldsymbol{\Psi}_{4,1} + \boldsymbol{\Psi}_5(\tau_1) + \boldsymbol{\Psi}_6(d) & \mathbf{H}_1^T \mathbf{P}_2 & \mathbf{H}_3^T \mathbf{P}_4 \\ & -(\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_4) & \mathbf{0} \\ & * & -(\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_8) \end{pmatrix} < \mathbf{0}, \quad (14)$$

$$\begin{pmatrix} \boldsymbol{\Psi}_1 + \boldsymbol{\Psi}_2 + \boldsymbol{\Psi}_{3,1} + \boldsymbol{\Psi}_{4,2} + \boldsymbol{\Psi}_5(\tau_1) + \boldsymbol{\Psi}_6(h) & \mathbf{H}_1^T \mathbf{P}_2 & \mathbf{H}_4^T \mathbf{P}_3^T \\ & -(\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_4) & \mathbf{0} \\ & * & -(\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_7) \end{pmatrix} < \mathbf{0}, \quad (15)$$

$$\begin{pmatrix} \boldsymbol{\Psi}_1 + \boldsymbol{\Psi}_2 + \boldsymbol{\Psi}_{3,2} + \boldsymbol{\Psi}_{4,1} + \boldsymbol{\Psi}_5(\tau_2) + \boldsymbol{\Psi}_6(d) & \mathbf{H}_2^T \mathbf{P}_1^T & \mathbf{H}_3^T \mathbf{P}_4 \\ & -(\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_3) & \mathbf{0} \\ & * & -(\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_8) \end{pmatrix} < \mathbf{0}, \quad (16)$$

$$\begin{pmatrix} \boldsymbol{\Psi}_1 + \boldsymbol{\Psi}_2 + \boldsymbol{\Psi}_{3,2} + \boldsymbol{\Psi}_{4,2} + \boldsymbol{\Psi}_5(\tau_2) + \boldsymbol{\Psi}_6(h) & \mathbf{H}_2^T \mathbf{P}_1^T & \mathbf{H}_4^T \mathbf{P}_3^T \\ & -(\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_3) & \mathbf{0} \\ & * & -(\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_7) \end{pmatrix} < \mathbf{0}. \quad (17)$$

$\mathbf{0}$ 元素除外, $\boldsymbol{\Psi}_1 = [\boldsymbol{\Psi}_{ij}]_{27 \times 27}$ 的剩余元素如下:

$$\begin{aligned} \boldsymbol{\Psi}_{11} &= \mathbf{Q}_1 + \mathbf{R}_1 + \tau_1^2 \mathbf{S}_1 + (\tau_2 - \tau_1)^2 \mathbf{S}_2 + d^2 \mathbf{S}_3 + (h - d)^2 \mathbf{S}_4 - 9\mathbf{T}_1 - 9\mathbf{T}_3 - 6\mathbf{J}_1 - 6\mathbf{J}_5 - \\ & \mathbf{D}^T \mathbf{F} \mathbf{A}_1 \mathbf{D} + \text{Sym}(\mathbf{G}^T \mathbf{A}) - \frac{\pi^2}{4} \mathbf{U}, \end{aligned}$$

$$\Psi_{12} = 3T_1, \Psi_{13} = G^T B, \Psi_{15} = 3T_3, \Psi_{16} = -YC + A^T G + \frac{\pi^2}{4} U,$$

$$\Psi_{18} = YC - A^T G, \Psi_{19} = D^T [W^+ \lambda - W^- \gamma] D + A^T G - G^T,$$

$$\Psi_{1,10} = D^T A_2 F + G^T E, \Psi_{1,11} = -24T_1 - 6J_1, \Psi_{1,14} = -24T_3 + 6J_5, \Psi_{1,17} = 30T_1 + 12J_1 + G^T,$$

$$\Psi_{1,20} = 30T_3 + 12J_5, \Psi_{22} = Q_2 - Q_1 - 9T_1 - 6J_2 - 6J_3, \Psi_{2,11} = 36T_1 + 18J_2, \Psi_{2,12} = -6J_3,$$

$$\Psi_{2,17} = -30T_1 - 12J_2, \Psi_{2,18} = 12J_3, \Psi_{33} = (1 - u_1)Q_3 + (u_2 - 1)Q_2 - 6J_3 - 6J_4, \Psi_{36} = B^T G,$$

$$\Psi_{38} = -B^T G, \Psi_{39} = B^T G, \Psi_{3,12} = 18J_4, \Psi_{3,13} = -6J_3, \Psi_{3,18} = -12J_4, \Psi_{3,19} = 12J_3,$$

$$\Psi_{44} = -Q_3 - 6J_4, \Psi_{4,13} = 18J_4, \Psi_{4,19} = -12J_4, \Psi_{55} = -9T_3 - 6J_6 - 6J_7 + R_2 - R_1,$$

$$\Psi_{5,14} = 36T_3 + 18J_6, \Psi_{5,15} = -6J_7, \Psi_{5,20} = -30T_3 - 12J_6, \Psi_{5,21} = 12J_7,$$

$$\Psi_{66} = (1 + \kappa_1 \mu_1) \Theta_3 + (1 + \kappa_2 \mu_2) \Theta_2 - 6J_7 - 6J_8 - \text{Sym}(YC) - \frac{\pi^2}{4} U,$$

$$\Psi_{68} = -(1 + \kappa_1 \mu_1) \Theta_3 + \text{Sym}(YC), \Psi_{69} = -C^T Y^T - G^T, \Psi_{6,10} = G^T E, \Psi_{6,15} = 18J_8,$$

$$\Psi_{6,16} = -6J_7, \Psi_{621} = -12J_8, \Psi_{622} = 12J_7, \Psi_{77} = -R_3 - 6J_8, \Psi_{7,16} = 18J_8, \Psi_{7,22} = -12J_8,$$

$$\Psi_{88} = -(\mu_1 + \mu_2) \Theta_1 + (1 + \kappa_1 \mu_1) \Theta_3 - \text{Sym}(YC), \Psi_{89} = C^T Y^T + G^T, \Psi_{8,10} = -G^T E,$$

$$\Psi_{99} = \tau_1^2 T_1 + (\tau_2 - \tau_1)^2 T_2 + d^2 T_3 + (h - d)^2 T_4 + \frac{\tau_1^2}{2} (J_1 + J_2) + \frac{(\tau_2 - \tau_1)^2}{2} (J_3 + J_4) +$$

$$\frac{d^2}{2} (J_5 + J_6) + \frac{(h - d)^2}{2} (J_7 + J_8) - \text{Sym}(G^T) + Q_4 + R_4 + h^2 U,$$

$$\Psi_{9,10} = D^T (Y - \lambda) + G^T E, \Psi_{10,10} = -F, \Psi_{11,11} = -4\tau_1^2 S_1 - 192T_1 - 66J_2 - 18J_1,$$

$$\Psi_{11,17} = 3\tau_1^2 S_1 + 180T_1 + 48J_2 + 24J_1, \Psi_{12,12} = -18J_3 - 66J_4, \Psi_{12,18} = 24J_3 + 48J_4,$$

$$\Psi_{13,13} = -18J_3 - 66J_4, \Psi_{13,19} = 24J_3 + 48J_4, \Psi_{14,14} = -4d^2 S_3 - 192T_3 - 18J_5 - 66J_6,$$

$$\Psi_{14,20} = 3d^2 S_3 + 180T_3 + 24J_5 + 48J_6, \Psi_{15,15} = -18J_7 - 66J_8, \Psi_{15,21} = 24J_7 + 48J_8,$$

$$\Psi_{16,16} = -18J_7 - 66J_8, \Psi_{16,22} = 24J_7 + 48J_8, \Psi_{17,17} = -3\tau_1 S_1 - 180T_1 - 36J_2 - 36J_1,$$

$$\Psi_{18,18} = -36J_3 - 36J_4, \Psi_{19,19} = -36J_3 - 36J_4, \Psi_{20,20} = -3d^2 S_3 - 180T_3 - 36J_5 - 36J_6,$$

$$\Psi_{21,21} = -36J_7 - 36J_8, \Psi_{22,22} = -36J_7 - 36J_8, \Psi_{23,23} = Q_5 - Q_4,$$

$$\Psi_{24,24} = (1 - u_1)Q_6 + (u_2 - 1)Q_5, \Psi_{25,25} = -Q_6, \Psi_{26,26} = R_5 - R_4, \Psi_{27,27} = -R_6,$$

$$\Psi_2 = \text{Sym}(L_1 PL),$$

$$\Psi_3(\tau(t)) = -H_1^T(\bar{T}_2 + \bar{J}_3)H_1 - H_2^T(\bar{T}_2 + \bar{J}_4)H_2 + H_1^T \bar{J}_3 H_1 + H_2^T \bar{J}_4 H_2 -$$

$$\frac{\tau_2 - \tau(t)}{\tau_2 - \tau_1} \begin{pmatrix} H_1 \\ H_2 \end{pmatrix}^T \begin{pmatrix} \bar{T}_2 + \bar{J}_3 - P_2(\bar{T}_2 + \bar{J}_4)^{-1} P_2^T & P_1 \\ * & \mathbf{0} \end{pmatrix} \begin{pmatrix} H_1 \\ H_2 \end{pmatrix} -$$

$$\frac{\tau(t) - \tau_1}{\tau_2 - \tau_1} \begin{pmatrix} H_1 \\ H_2 \end{pmatrix}^T \begin{pmatrix} \mathbf{0} & P_2 \\ * & \bar{T}_2 + \bar{J}_4 - P_1^T(\bar{T}_2 + \bar{J}_3)^{-1} P_1 \end{pmatrix} \begin{pmatrix} H_1 \\ H_2 \end{pmatrix},$$

$$\Psi_4(h(t)) = -H_3^T(\bar{T}_4 + \bar{J}_7)H_3 - H_4^T(\bar{T}_4 + \bar{J}_8)H_4 + H_3^T \bar{J}_7 H_3 + H_4^T \bar{J}_8 H_4 -$$

$$\frac{h - h(t)}{h - d} \begin{pmatrix} H_3 \\ H_4 \end{pmatrix}^T \begin{pmatrix} \bar{T}_4 + \bar{J}_7 - P_4(\bar{T}_4 + \bar{J}_8)^{-1} P_4^T & P_3 \\ * & \mathbf{0} \end{pmatrix} \begin{pmatrix} H_3 \\ H_4 \end{pmatrix} -$$

$$\frac{h(t) - d}{h - d} \begin{pmatrix} H_3 \\ H_4 \end{pmatrix}^T \begin{pmatrix} \mathbf{0} & P_4 \\ * & \bar{T}_4 + \bar{J}_8 - P_3^T(\bar{T}_4 + \bar{J}_7)^{-1} P_3 \end{pmatrix} \begin{pmatrix} H_3 \\ H_4 \end{pmatrix},$$

$$\Psi_5(\tau(t)) = -(\tau_2 - \tau_1)[\tau(t) - \tau_1][e_{12}^T S_2 e_{12} - 3(-e_{12} + e_{18})^T S_2(-e_{12} + e_{18}) + \text{Sym}(L_2 PL)] -$$

$$(\tau_2 - \tau_1)[\tau_2 - \tau(t)][e_{13}^T S_2 e_{13} - 3(-e_{13} + e_{19})^T S_2(-e_{13} + e_{19}) + \text{Sym}(L_3 PL)],$$

$$\Psi_6(h(t)) = -(h - d)[h(t) - d][e_{15}^T S_4 e_{15} - 3(-e_{15} + e_{21})^T S_4(-e_{15} + e_{21}) + \text{Sym}(L_4 PL)] -$$

$$(h-d)[h-h(t)][e_{16}^T S_4 e_{16} - 3(-e_{16} + e_{22})^T S_4 (-e_{16} + e_{22}) + \text{Sym}(L_5 PL)],$$

$$\Psi_{3,1} = \begin{pmatrix} 2\bar{T}_2 + \bar{J}_3 & P_1 \\ * & \bar{T}_2 \end{pmatrix}, \Psi_{3,2} = \begin{pmatrix} \bar{T}_2 & P_2 \\ * & 2\bar{T}_2 + \bar{J}_4 \end{pmatrix}, \Psi_{4,1} = \begin{pmatrix} 2\bar{T}_4 + \bar{J}_7 & P_3 \\ * & \bar{T}_4 \end{pmatrix}, \Psi_{4,2} = \begin{pmatrix} \bar{T}_4 & P_4 \\ * & 2\bar{T}_4 + \bar{J}_8 \end{pmatrix}.$$

则误差系统(6)是全局渐近稳定的,且时滞反馈控制增益矩阵为

$$K = G^{-1}Y.$$

证明 构造的 Lyapunov-Krasovskii 泛函如下:

$$V(t) = \sum_{i=1}^7 V_i(t), \quad (18)$$

$$V_1(t) = \eta^T(t)P\eta(t),$$

$$V_2(t) = \int_{t-\tau_1}^t e^T(s)Q_1 e(s) ds + \int_{t-\tau(t)}^{t-\tau_1} e^T(s)Q_2 e(s) ds + \int_{t-\tau_2}^{t-\tau(t)} e^T(s)Q_3 e(s) ds + \\ \int_{t-\tau_1}^t \dot{e}^T(s)Q_4 \dot{e}(s) ds + \int_{t-\tau(t)}^{t-\tau_1} \dot{e}^T(s)Q_5 \dot{e}(s) ds + \int_{t-\tau_2}^{t-\tau(t)} \dot{e}^T(s)Q_6 \dot{e}(s) ds + \\ \int_{t-d}^t e^T(s)R_1 e(s) ds + \int_{t-h(t)}^{t-d} e^T(s)R_2 e(s) ds + \int_{t-h}^{t-h(t)} e^T(s)R_3 e(s) ds + \\ \int_{t-d}^t \dot{e}^T(s)R_4 \dot{e}(s) ds + \int_{t-h(t)}^{t-d} \dot{e}^T(s)R_5 \dot{e}(s) ds + \int_{t-h}^{t-h(t)} \dot{e}^T(s)R_6 \dot{e}(s) ds,$$

$$V_3(t) = \tau_1 \int_{t-\tau_1}^t \int_{\theta}^t e^T(s)S_1 e(s) ds d\theta + \tau_1 \int_{t-\tau_1}^t \int_{\theta}^t \dot{e}^T(s)T_1 \dot{e}(s) ds d\theta + \\ (\tau_2 - \tau_1) \int_{t-\tau_2}^{t-\tau_1} \int_{\theta}^t e^T(s)S_2 e(s) ds d\theta + (\tau_2 - \tau_1) \int_{t-\tau_2}^{t-\tau_1} \int_{\theta}^t \dot{e}^T(s)T_2 \dot{e}(s) ds d\theta + \\ d \int_{t-d}^t \int_{\theta}^t e^T(s)S_3 e(s) ds d\theta + d \int_{t-d}^t \int_{\theta}^t \dot{e}^T(s)T_3 \dot{e}(s) ds d\theta + \\ (h-d) \int_{t-h}^{t-d} \int_{\theta}^t e^T(s)S_4 e(s) ds d\theta + (h-d) \int_{t-h}^{t-d} \int_{\theta}^t \dot{e}^T(s)T_4 \dot{e}(s) ds d\theta,$$

$$V_4(t) = \int_{t-\tau_1}^t \int_{\rho}^t \int_{\theta}^t \dot{e}^T(s)J_1 \dot{e}(s) ds d\theta d\rho + \int_{t-\tau_1}^t \int_{t-\tau_1}^{\rho} \int_{\theta}^t \dot{e}^T(s)J_2 \dot{e}(s) ds d\theta d\rho + \\ \int_{t-\tau_2}^{t-\tau_1} \int_{\rho}^t \int_{\theta}^t \dot{e}^T(s)J_3 \dot{e}(s) ds d\theta d\rho + \int_{t-\tau_2}^{t-\tau_1} \int_{t-\tau_2}^{\rho} \int_{\theta}^t \dot{e}^T(s)J_4 \dot{e}(s) ds d\theta d\rho + \\ \int_{t-d}^t \int_{\rho}^t \int_{\theta}^t \dot{e}^T(s)J_5 \dot{e}(s) ds d\theta d\rho + \int_{t-d}^t \int_{t-d}^{\rho} \int_{\theta}^t \dot{e}^T(s)J_6 \dot{e}(s) ds d\theta d\rho + \\ \int_{t-h}^{t-d} \int_{\rho}^t \int_{\theta}^t \dot{e}^T(s)J_7 \dot{e}(s) ds d\theta d\rho + \int_{t-h}^{t-d} \int_{t-h}^{\rho} \int_{\theta}^t \dot{e}^T(s)J_8 \dot{e}(s) ds d\theta d\rho,$$

$$V_5(t) = h^2 \int_{t_s}^t \dot{e}(s)U\dot{e}(s) ds - \frac{\pi^2}{4} \int_{t_s}^t [e(s) - e(t_k)]^T U [e(s) - e(t_k)] ds,$$

$$V_6(t) = 2 \sum_{i=1}^m \left\{ \int_0^{d_i^T e} [\lambda_i(\omega_i^+ s - g_i(s))] ds + \int_0^{d_i^T e} [\gamma_i(g_i(s) - \omega_i^- s)] ds \right\},$$

$$V_7(t) = \frac{1}{2} \sigma_1^2(t) + \frac{1}{2} \sigma_2^2(t).$$

沿着误差系统式(6)的轨线,求 $V(t)$ 关于时间的导数:

$$\dot{V}_1(t) = 2\eta^T(t)P\dot{\eta}(t), \quad (19)$$

$$\dot{V}_2(t) = e^T(t)(Q_1 + R_1)e(t) + e^T(t-\tau_1)(Q_2 - Q_1)e(t-\tau_1) - e^T(t-\tau_2)Q_3 e(t-\tau_2) + \\ (1 - \dot{\tau}(t))\dot{e}^T(t-\tau(t))(Q_5 - Q_6)\dot{e}(t-\tau(t)) + \dot{e}^T(t)Q_4 \dot{x}(t) + \\ \dot{e}^T(t-\tau_1)(Q_5 - Q_6)\dot{e}(t-\tau_1) - \dot{e}^T(t-\tau_2)Q_6 \dot{e}(t-\tau_2) + \\ e^T(t-d)(R_2 - R_1)e(t-d) - e^T(t-h)R_3 e(t-h) + \dot{e}^T(t)R_4 \dot{e}(t) + \\ \dot{e}^T(t-d)(R_5 - R_4)\dot{e}(t-d) - \dot{e}^T(t-h)R_6 \dot{e}(t-h), \quad (20)$$

$$\begin{aligned} \dot{V}_3(t) = & \mathbf{e}^T(t) [\tau_1^2 \mathbf{S}_1 + (\tau_2 - \tau_1)^2 \mathbf{S}_2 + d^2 \mathbf{S}_3 + (h-d)^2 \mathbf{S}_4] \mathbf{e}(t) + \\ & \dot{\mathbf{e}}^T(t) (\tau_1^2 \mathbf{T}_1 + (\tau_2 - \tau_1)^2 \mathbf{T}_2 + d^2 \mathbf{T}_3 + (h-d)^2 \mathbf{T}_4) \dot{\mathbf{e}}(t) - \\ & \tau_1 \int_{t-\tau_1}^t \mathbf{e}^T(s) \mathbf{S}_1 \mathbf{e}(s) ds - (\tau_2 - \tau_1) \int_{t-\tau_2}^{t-\tau_1} \mathbf{e}^T(s) \mathbf{S}_2 \mathbf{e}(s) ds - \tau_1 \int_{t-\tau_1}^t \dot{\mathbf{e}}^T(s) \mathbf{T}_1 \dot{\mathbf{e}}(s) ds - \\ & (\tau_2 - \tau_1) \int_{t-\tau_2}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{T}_2 \dot{\mathbf{e}}(s) ds - d \int_{t-d}^t \mathbf{e}^T(s) \mathbf{S}_3 \mathbf{e}(s) ds - (h-d) \int_{t-h}^{t-d} \mathbf{e}^T(s) \mathbf{S}_4 \mathbf{e}(s) ds - \\ & d \int_{t-d}^t \dot{\mathbf{e}}^T(s) \mathbf{T}_3 \dot{\mathbf{e}}(s) ds - (h-d) \int_{t-h}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{T}_4 \dot{\mathbf{e}}(s) ds, \end{aligned} \quad (21)$$

$$\begin{aligned} \dot{V}_4(t) = & \frac{\tau_1^2}{2} \dot{\mathbf{e}}^T(t) \left[\frac{\tau_1^2}{2} (\mathbf{J}_1 + \mathbf{J}_2) + \frac{(\tau_2 - \tau_1)^2}{2} (\mathbf{J}_3 + \mathbf{J}_4) + \frac{d^2}{2} (\mathbf{J}_5 + \mathbf{J}_6) + \right. \\ & \left. \frac{(h-d)^2}{2} (\mathbf{J}_7 + \mathbf{J}_8) \right] \dot{\mathbf{e}}(t) - \\ & \int_{t-\tau_1}^t \int_{\theta}^t \dot{\mathbf{e}}^T(s) \mathbf{J}_1 \dot{\mathbf{e}}(s) ds d\theta - \int_{t-\tau_1}^t \int_{t-\tau_1}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_2 \dot{\mathbf{e}}(s) ds d\theta - \int_{t-\tau_2}^{t-\tau_1} \int_{t-\tau_2}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_3 \dot{\mathbf{e}}(s) ds - \\ & \int_{t-\tau_2}^{t-\tau_1} \int_{t-\tau_2}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_4 \dot{\mathbf{e}}(s) ds d\theta - \int_{t-d}^t \int_{\theta}^t \dot{\mathbf{e}}^T(s) \mathbf{J}_5 \dot{\mathbf{e}}(s) ds d\theta - \int_{t-d}^t \int_{t-d}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_6 \dot{\mathbf{e}}(s) ds d\theta - \\ & \int_{t-h}^{t-d} \int_{\theta}^t \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) ds d\theta - \int_{t-h}^{t-d} \int_{t-h}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) ds d\theta, \end{aligned} \quad (22)$$

$$\dot{V}_5(t) = h^2 \dot{\mathbf{e}}^T(t) \mathbf{U} \dot{\mathbf{e}}(t) - \frac{\pi^2}{4} [\mathbf{e}(t) - \mathbf{e}(t_s)]^T \mathbf{U} [\mathbf{e}(t) - \mathbf{e}(t_s)], \quad (23)$$

$$\dot{V}_6(t) = 2 \{ [\mathbf{W}^+ \mathbf{D} \mathbf{e}(t) - \mathbf{g}(\mathbf{D} \mathbf{e}(t))]^T \boldsymbol{\lambda} \mathbf{D} \dot{\mathbf{e}}(t) + [\mathbf{g}(\mathbf{D} \mathbf{e}(t) - \mathbf{W}^- \mathbf{D} \mathbf{e}(t)) \boldsymbol{\gamma} \mathbf{D} \dot{\mathbf{e}}(t)] \}, \quad (24)$$

$$\begin{aligned} \dot{V}_7(t) = & \sigma_1(t) \dot{\sigma}_1(t) + \sigma_2(t) \dot{\sigma}_2(t) = \\ & \left[\frac{1}{\sigma_1(t)} - \mu_1 \right] [\boldsymbol{\varepsilon}^T(t_s) \boldsymbol{\Theta}_1 \boldsymbol{\varepsilon}(t_s) - \kappa_1 \mathbf{e}^T(t_k) \boldsymbol{\Theta}_3 \mathbf{e}(t_k)] + \\ & \left[\frac{1}{\sigma_2(t)} - \mu_2 \right] [\boldsymbol{\varepsilon}^T(t_s) \boldsymbol{\Theta}_1 \boldsymbol{\varepsilon}(t_s) - \kappa_2 \mathbf{e}^T(t_s) \boldsymbol{\Theta}_2 \mathbf{e}(t_s)]. \end{aligned} \quad (25)$$

由引理2可得

$$- \int_{t-\tau_1}^t \mathbf{e}^T(s) \mathbf{S}_1 \mathbf{e}(s) ds \leq -\tau_1 \boldsymbol{\chi}^T(t) \mathbf{S}_1 \boldsymbol{\chi}(t) - 3\tau_1 [-\boldsymbol{\chi}(t) + \boldsymbol{\kappa}(t)]^T \mathbf{S}_1 [-\boldsymbol{\chi}(t) + \boldsymbol{\kappa}(t)], \quad (26)$$

$$\begin{aligned} - \int_{t-\tau_2}^{t-\tau_1} \mathbf{e}^T(s) \mathbf{S}_2 \mathbf{e}(s) ds = & - \int_{t-\tau(t)}^{t-\tau_1} \mathbf{e}^T(s) \mathbf{S}_2 \mathbf{e}(s) ds - \int_{t-\tau_2}^{t-\tau(t)} \mathbf{e}^T(s) \mathbf{S}_2 \mathbf{e}(s) ds \leq \\ & - (\tau(t) - \tau_1) \boldsymbol{\delta}^T(t) \mathbf{S}_2 \boldsymbol{\delta}(t) - 3(\tau(t) - \tau_1) [-\boldsymbol{\delta}(t) + \boldsymbol{\lambda}(t)]^T \mathbf{S}_2 [-\boldsymbol{\delta}(t) + \boldsymbol{\lambda}(t)] - \\ & (\tau_2 - \tau(t)) \boldsymbol{\varepsilon}^T(t) \mathbf{S}_2 \boldsymbol{\varepsilon}(t) - 3(\tau_2 - \tau(t)) [-\boldsymbol{\varepsilon}(t) + \boldsymbol{\nu}(t)]^T \mathbf{S}_2 [-\boldsymbol{\varepsilon}(t) + \boldsymbol{\nu}(t)], \end{aligned} \quad (27)$$

$$- \int_{t-d}^t \mathbf{e}^T(s) \mathbf{S}_3 \mathbf{e}(s) ds \leq -d \boldsymbol{\phi}^T(t) \mathbf{S}_3 \boldsymbol{\phi}(t) - 3d [\boldsymbol{\phi}(t) - \boldsymbol{\omega}(t)]^T \mathbf{S}_3 [\boldsymbol{\phi}(t) - \boldsymbol{\omega}(t)], \quad (28)$$

$$\begin{aligned} - \int_{t-h}^{t-d} \mathbf{e}^T(s) \mathbf{S}_4 \mathbf{e}(s) ds = & - \int_{t-h(t)}^{t-d} \mathbf{e}^T(s) \mathbf{S}_4 \mathbf{e}(s) ds - \int_{t-h}^{t-h(t)} \mathbf{e}^T(s) \mathbf{S}_4 \mathbf{e}(s) ds \leq \\ & - (h(t) - d) \boldsymbol{\varphi}^T(t) \mathbf{S}_4 \boldsymbol{\varphi}(t) - 3(h(t) - d) [-\boldsymbol{\varphi}(t) + \boldsymbol{\theta}(t)]^T \mathbf{S}_4 [-\boldsymbol{\varphi}(t) + \boldsymbol{\theta}(t)] - \\ & (h - h(t)) \boldsymbol{\gamma}^T(t) \mathbf{S}_4 \boldsymbol{\gamma}(t) - 3(h - h(t)) [-\boldsymbol{\varphi}(t) + \boldsymbol{\vartheta}(t)]^T \mathbf{S}_4 [-\boldsymbol{\varphi}(t) + \boldsymbol{\vartheta}(t)]. \end{aligned} \quad (29)$$

由引理3可得

$$\begin{aligned} - \tau_1 \int_{t-\tau_1}^t \dot{\mathbf{e}}^T(s) \mathbf{T}_1 \dot{\mathbf{e}}(s) ds \leq & - [\mathbf{e}(t) - \mathbf{e}(t - \tau_1)]^T \mathbf{T}_1 [\mathbf{e}(t) - \mathbf{e}(t - \tau_1)] - \\ & 3 [\mathbf{e}(t) + \mathbf{e}(t - \tau_1) - 2\boldsymbol{\chi}(t)]^T \mathbf{T}_1 [\mathbf{e}(t) + \mathbf{e}(t - \tau_1) - 2\boldsymbol{\chi}(t)] - \\ & 5 [\mathbf{e}(t) - \mathbf{e}(t - \tau_1) + 6\boldsymbol{\chi}(t) - 6\boldsymbol{\kappa}(t)]^T \mathbf{T}_1 [\mathbf{e}(t) - \mathbf{e}(t - \tau_1) + 6\boldsymbol{\chi}(t) - 6\boldsymbol{\kappa}(t)], \end{aligned} \quad (30)$$

$$- d \int_{t-d}^t \dot{\mathbf{e}}^T(s) \mathbf{T}_3 \dot{\mathbf{e}}(s) ds \leq - [\mathbf{e}(t) - \mathbf{e}(t - d)]^T \mathbf{T}_3 [\mathbf{e}(t) - \mathbf{e}(t - d)] -$$

$$\begin{aligned} & [\mathbf{e}(t) + \mathbf{e}(t-d) - 2\boldsymbol{\phi}(t)]^T \mathbf{T}_3 [\mathbf{e}(t) + \mathbf{e}(t-d) - 2\boldsymbol{\phi}(t)] - \\ & [\mathbf{e}(t) - \mathbf{e}(t-d) + 6\boldsymbol{\phi}(t) - 6\boldsymbol{\omega}(t)]^T \mathbf{T}_3 [\mathbf{e}(t) - \mathbf{e}(t-d) + 6\boldsymbol{\phi}(t) - 6\boldsymbol{\omega}(t)], \end{aligned} \quad (31)$$

$$\begin{aligned} & - (\tau_2 - \tau(t)) \int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{J}_3 \dot{\mathbf{e}}(s) ds - (\tau(t) - \tau_1) \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_4 \dot{\mathbf{e}}(s) ds = \\ & - \frac{1-\rho_1}{\rho_1} [\tau(t) - \tau_1] \int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{J}_3 \dot{\mathbf{e}}(s) ds - \frac{1-\rho_2}{\rho_2} [\tau_2 - \tau(t)] \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_4 \dot{\mathbf{e}}(s) ds, \end{aligned} \quad (32)$$

$$\begin{aligned} & - (\tau_2 - \tau_1) \int_{t-\tau_2}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{T}_2 \dot{\mathbf{e}}(s) ds = \\ & - (\tau_2 - \tau_1) \left[\int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{T}_2 \dot{\mathbf{e}}(s) ds + \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) \mathbf{T}_2 \dot{\mathbf{e}}(s) ds \right] = \\ & - \frac{1}{\rho_1} (\tau(t) - \tau_1) \int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{T}_2 \dot{\mathbf{e}}(s) ds - \frac{1}{\rho_2} (\tau_2 - \tau(t)) \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) \mathbf{T}_2 \dot{\mathbf{e}}(s) ds. \end{aligned} \quad (33)$$

由引理 4 和引理 1 可得

$$\begin{aligned} & - \frac{1}{\rho_1} (\tau(t) - \tau_1) \int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{e}}^T(s) (\mathbf{T}_2 + \mathbf{J}_3) \dot{\mathbf{e}}(s) ds - \frac{1}{\rho_2} (\tau_2 - \tau(t)) \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) (\mathbf{T}_2 + \mathbf{J}_4) \dot{\mathbf{e}}(s) ds + \\ & (\tau(t) - \tau_1) \int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{e}}^T(s) (\mathbf{T}_2 + \mathbf{J}_3) \dot{\mathbf{e}}(s) ds + (\tau_2 - \tau(t)) \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) (\mathbf{T}_2 + \mathbf{J}_4) \dot{\mathbf{e}}(s) ds \leq \\ & \boldsymbol{\xi}^T(t) [-\mathbf{H}_1^T (\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_3) \mathbf{H}_1 - \mathbf{H}_2^T (\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_4) \mathbf{H}_2 + \mathbf{H}_1^T \bar{\mathbf{J}}_3 \mathbf{H}_1 + \mathbf{H}_2^T \bar{\mathbf{J}}_4 \mathbf{H}_2] \boldsymbol{\xi}(t) + \\ & \boldsymbol{\xi}^T(t) \left[-\rho_2 \begin{pmatrix} \mathbf{H}_1 \\ \mathbf{H}_2 \end{pmatrix}^T \begin{pmatrix} \bar{\mathbf{T}} + \bar{\mathbf{J}}_3 - \mathbf{P}_2 (\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_4)^{-1} \mathbf{P}_2^T & \mathbf{P}_1 \\ * & \mathbf{0} \end{pmatrix} \begin{pmatrix} \mathbf{H}_1 \\ \mathbf{H}_2 \end{pmatrix} \right] \boldsymbol{\xi}(t) + \\ & \boldsymbol{\xi}^T(t) \left[-\rho_1 \begin{pmatrix} \mathbf{H}_1 \\ \mathbf{H}_2 \end{pmatrix}^T \begin{pmatrix} \mathbf{0} & \mathbf{P}_2 \\ * & \bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_4 - \mathbf{P}_1^T (\bar{\mathbf{T}}_2 + \bar{\mathbf{J}}_3)^{-1} \mathbf{P}_1 \end{pmatrix} \begin{pmatrix} \mathbf{H}_1 \\ \mathbf{H}_2 \end{pmatrix} \right] \boldsymbol{\xi}(t), \end{aligned} \quad (34)$$

$$\begin{aligned} & - \int_{t-\tau_1}^t \int_{\theta}^t \dot{\mathbf{e}}^T(s) \mathbf{J}_1 \dot{\mathbf{e}}(s) ds d\theta \leq - [\mathbf{e}(t) - \boldsymbol{\chi}_1(t)]^T \mathbf{J}_1 [\mathbf{e}(t) - \boldsymbol{\chi}_1(t)] - \\ & 2 [\mathbf{e}(t) + 2\boldsymbol{\chi}_1(t) - 3\boldsymbol{\kappa}(t)]^T \mathbf{J}_1 [\mathbf{e}(t) + 2\boldsymbol{\chi}_1(t) - 3\boldsymbol{\kappa}(t)], \end{aligned} \quad (35)$$

$$\begin{aligned} & - \int_{t-\tau_1}^t \int_{t-\tau_1}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_2 \dot{\mathbf{e}}(s) ds d\theta \leq - [\mathbf{e}(t - \tau_1) + \boldsymbol{\chi}_1(t)]^T \mathbf{J}_2 [\mathbf{e}(t - \tau_1) + \boldsymbol{\chi}_1(t)] - \\ & 2 [\mathbf{e}(t - \tau_1) - 4\boldsymbol{\chi}_1(t) + 3\boldsymbol{\kappa}(t)]^T \mathbf{J}_2 [\mathbf{e}(t - \tau_1) - 4\boldsymbol{\chi}_1(t) + 3\boldsymbol{\kappa}(t)], \end{aligned} \quad (36)$$

$$\begin{aligned} & - \int_{t-\tau_2}^{t-\tau_1} \int_{\theta}^{t-\tau_1} \dot{\mathbf{e}}^T(s) \mathbf{J}_3 \dot{\mathbf{e}}(s) ds d\theta = - \left(\int_{t-\tau(t)}^{t-\tau_1} \int_{\theta}^{t-\tau_1} + \int_{t-\tau_2}^{t-\tau(t)} \int_{t-\tau(t)}^{t-\tau_1} + \int_{t-\tau_2}^{t-\tau(t)} \int_{\theta}^{t-\tau(t)} \right) \dot{\mathbf{e}}^T(s) \mathbf{J}_3 \dot{\mathbf{e}}(s) ds d\theta \leq \\ & - 2 [\mathbf{e}(t - \tau_1) - \boldsymbol{\delta}(t)]^T \mathbf{J}_3 [\mathbf{e}(t - \tau_1) - \boldsymbol{\delta}(t)] - \\ & 4 [\mathbf{e}(t - \tau_1) + 2\boldsymbol{\delta}(t) - 3\boldsymbol{\lambda}(t)]^T \mathbf{J}_3 [\mathbf{e}(t - \tau_1) + 2\boldsymbol{\delta}(t) - 3\boldsymbol{\lambda}(t)] - \\ & (\tau_2 - \tau(t)) \int_{t-\tau(t)}^{t-\tau_1} \dot{\mathbf{x}}^T(s) \mathbf{J}_3 \dot{\mathbf{x}}(s) ds - \\ & 2 [\mathbf{e}(t - \tau(t)) - \boldsymbol{\varepsilon}(t)]^T \mathbf{J}_3 [\mathbf{e}(t - \tau(t)) - \boldsymbol{\varepsilon}(t)] - \\ & 4 [\mathbf{e}(t - \tau(t)) + 2\boldsymbol{\varepsilon}(t) - 3\boldsymbol{\nu}(t)]^T \mathbf{J}_3 [\mathbf{e}(t - \tau(t)) + 2\boldsymbol{\varepsilon}(t) - 3\boldsymbol{\nu}(t)], \end{aligned} \quad (37)$$

$$\begin{aligned} & - \int_{t-\tau_2}^{t-\tau_1} \int_{t-\tau_2}^{\theta} \dot{\mathbf{x}}^T(s) \mathbf{J}_4 \dot{\mathbf{x}}(s) ds d\theta = - \left(\int_{t-\tau(t)}^{t-\tau_1} \int_{t-\tau_2}^{t-\tau(t)} + \int_{t-\tau(t)}^{t-\tau_1} \int_{t-\tau(t)}^{\theta} + \int_{t-\tau_2}^{t-\tau(t)} \int_{t-\tau_2}^{\theta} \right) \dot{\mathbf{e}}^T(s) \mathbf{J}_4 \dot{\mathbf{e}}(s) ds d\theta \leq \\ & - (\tau(t) - \tau_1) \int_{t-\tau_2}^{t-\tau(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_4 \dot{\mathbf{e}}(s) ds - \\ & 2 [\mathbf{e}(t - \tau(t)) - \boldsymbol{\delta}(t)]^T \mathbf{J}_4 [\mathbf{e}(t - \tau(t)) - \boldsymbol{\delta}(t)] - \\ & 4 [\mathbf{e}(t - \tau(t)) - 4\boldsymbol{\delta}(t) + 3\boldsymbol{\lambda}(t)]^T \mathbf{J}_4 [\mathbf{e}(t - \tau(t)) - 4\boldsymbol{\delta}(t) + 3\boldsymbol{\lambda}(t)] - \\ & 2 [-\mathbf{e}(t - \tau_2) + \boldsymbol{\varepsilon}(t)]^T \mathbf{J}_4 [-\mathbf{e}(t - \tau_2) + \boldsymbol{\varepsilon}(t)] - \\ & 4 [\mathbf{e}(t - \tau_2) - 4\boldsymbol{\varepsilon}(t) + 3\boldsymbol{\nu}(t)]^T \mathbf{J}_4 [\mathbf{e}(t - \tau_2) - 4\boldsymbol{\varepsilon}(t) + 3\boldsymbol{\nu}(t)], \end{aligned} \quad (38)$$

$$\begin{aligned}
& - \int_{t-d}^t \int_{\theta}^t \dot{\mathbf{e}}^T(s) \mathbf{J}_5 \dot{\mathbf{e}}(s) \, ds d\theta \leq - [\mathbf{e}(t) - \boldsymbol{\phi}(t)]^T \mathbf{J}_5 [\mathbf{e}(t) - \boldsymbol{\phi}(t)] - \\
& \quad 2 [-\mathbf{e}(t) - 2\boldsymbol{\phi}(t) + 3\boldsymbol{\omega}(t)]^T \mathbf{J}_5 [-\mathbf{e}(t) - 2\boldsymbol{\phi}(t) + 3\boldsymbol{\omega}(t)], \tag{39}
\end{aligned}$$

$$\begin{aligned}
& - \int_{t-d}^t \int_{t-d}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_6 \dot{\mathbf{e}}(s) \, ds d\theta \leq - [-\mathbf{e}(t-d) - \boldsymbol{\phi}(t)]^T \mathbf{J}_6 [-\mathbf{e}(t-d) - \boldsymbol{\phi}(t)] - \\
& \quad 2 [\mathbf{e}(t-d) - 4\boldsymbol{\phi}(t) + 3\boldsymbol{\omega}(t)]^T \mathbf{J}_6 [\mathbf{e}(t-d) - 4\boldsymbol{\phi}(t) + 3\boldsymbol{\omega}(t)], \tag{40}
\end{aligned}$$

$$\begin{aligned}
& - \int_{t-h}^{t-d} \int_{\theta}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) \, ds d\theta = - \left(\int_{t-h(t)}^{t-d} \int_{\theta}^{t-d} + \int_{t-h}^{t-h(t)} \int_{t-h(t)}^{t-d} + \int_{t-h}^{t-h(t)} \int_{\theta}^{t-h(t)} \right) \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) \, ds \leq \\
& \quad - 2 [\mathbf{e}(t-d) - \boldsymbol{\varphi}(t)]^T \mathbf{J}_7 [\mathbf{e}(t-d) - \boldsymbol{\varphi}(t)] - \\
& \quad 4 [\mathbf{e}(t-d) + 2\boldsymbol{\varphi}(t) - 3\boldsymbol{\theta}(t)]^T \mathbf{J}_7 [\mathbf{e}(t-d) + 2\boldsymbol{\varphi}(t) - 3\boldsymbol{\theta}(t)] - \\
& \quad [h - h(t)] \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) \, ds - \\
& \quad 2 [\mathbf{e}(t-h(t)) - \boldsymbol{\gamma}(t)]^T \mathbf{J}_7 [\mathbf{e}(t-h(t)) - \boldsymbol{\gamma}(t)] - \\
& \quad 4 [\mathbf{e}(t-h(t)) + 2\boldsymbol{\gamma}(t) - 3\boldsymbol{\vartheta}(t)]^T \mathbf{J}_7 [\mathbf{e}(t-h(t)) + 2\boldsymbol{\gamma}(t) - 3\boldsymbol{\vartheta}(t)], \tag{41}
\end{aligned}$$

$$\begin{aligned}
& - \int_{t-h}^{t-d} \int_{t-h}^{\theta} \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) \, ds d\theta = - \left(\int_{t-h(t)}^{t-d} \int_{t-h}^{t-h(t)} + \int_{t-h(t)}^{t-d} \int_{t-h(t)}^{\theta} + \int_{t-h}^{t-h(t)} \int_{t-h}^{\theta} \right) \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) \, ds \leq \\
& \quad - [h(t) - d] \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) \, ds - \\
& \quad 2 [-\mathbf{e}(t-h(t)) + \boldsymbol{\varphi}(t)]^T \mathbf{J}_8 [-\mathbf{e}(t-h(t)) + \boldsymbol{\varphi}(t)] - \\
& \quad 4 [\mathbf{e}(t-h(t)) - 4\boldsymbol{\varphi}(t) + 3\boldsymbol{\theta}(t)]^T \mathbf{J}_8 [\mathbf{e}(t-h(t)) - 4\boldsymbol{\varphi}(t) + 3\boldsymbol{\theta}(t)] - \\
& \quad 2 [-\mathbf{e}(t-h) + \boldsymbol{\gamma}(t)]^T \mathbf{J}_8 [-\mathbf{e}(t-h) + \boldsymbol{\gamma}(t)] - \\
& \quad 4 [\mathbf{e}(t-h) - 4\boldsymbol{\gamma}(t) + 3\boldsymbol{\vartheta}(t)]^T \mathbf{J}_8 [\mathbf{e}(t-h) - 4\boldsymbol{\gamma}(t) + 3\boldsymbol{\vartheta}(t)], \tag{42}
\end{aligned}$$

$$\begin{aligned}
& - [h - h(t)] \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) \, ds - [h(t) - d] \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) \, ds = \\
& \quad - \frac{1 - \rho_3}{\rho_3} [h(t) - d] \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) \, ds - \frac{1 - \rho_4}{\rho_4} [h - h(t)] \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) \, ds, \tag{43}
\end{aligned}$$

$$\begin{aligned}
& - (h - d) \int_{t-h}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{T}_4 \dot{\mathbf{e}}(s) \, ds = \\
& \quad - (h - d) \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{T}_4 \dot{\mathbf{e}}(s) \, ds - (h - d) \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) \mathbf{T}_4 \dot{\mathbf{e}}(s) \, ds = \\
& \quad - \frac{1}{\rho_3} [h(t) - d] \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{T}_4 \dot{\mathbf{e}}(s) \, ds - \frac{1}{\rho_4} [h - h(t)] \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) \mathbf{T}_4 \dot{\mathbf{e}}(s) \, ds, \tag{44}
\end{aligned}$$

$$\begin{aligned}
& - \frac{1}{\rho_3} [h(t) - d] \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) (\mathbf{T}_4 + \mathbf{J}_7) \dot{\mathbf{e}}(s) \, ds - \frac{1}{\rho_4} [h - h(t)] \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) (\mathbf{T}_4 + \mathbf{J}_8) \dot{\mathbf{e}}(s) \, ds + \\
& \quad [h(t) - d] \int_{t-h(t)}^{t-d} \dot{\mathbf{e}}^T(s) \mathbf{J}_7 \dot{\mathbf{e}}(s) \, ds + [h - h(t)] \int_{t-h}^{t-h(t)} \dot{\mathbf{e}}^T(s) \mathbf{J}_8 \dot{\mathbf{e}}(s) \, ds \leq \\
& \quad \boldsymbol{\xi}^T(t) [-\mathbf{H}_3^T (\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_7) \mathbf{H}_3 - \mathbf{H}_4^T (\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_8) \mathbf{H}_4 + \mathbf{H}_3^T \bar{\mathbf{J}}_7 \mathbf{H}_3 + \mathbf{H}_4^T \bar{\mathbf{J}}_8 \mathbf{H}_4] \boldsymbol{\xi}(t) + \\
& \quad \boldsymbol{\xi}^T(t) \begin{bmatrix} -\rho_4 \begin{pmatrix} \mathbf{H}_3 \\ \mathbf{H}_4 \end{pmatrix}^T \begin{pmatrix} \bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_7 - \mathbf{P}_4 (\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_8)^{-1} \mathbf{P}_4^T & \mathbf{P}_3 \\ * & \mathbf{0} \end{pmatrix} \begin{pmatrix} \mathbf{H}_3 \\ \mathbf{H}_4 \end{pmatrix} \\ \boldsymbol{\xi}^T(t) \begin{bmatrix} -\rho_3 \begin{pmatrix} \mathbf{H}_3 \\ \mathbf{H}_4 \end{pmatrix}^T \begin{pmatrix} \mathbf{0} & \mathbf{P}_4 \\ * & \bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_8 - \mathbf{P}_3^T (\bar{\mathbf{T}}_4 + \bar{\mathbf{J}}_7)^{-1} \mathbf{P}_3 \end{pmatrix} \begin{pmatrix} \mathbf{H}_3 \\ \mathbf{H}_4 \end{pmatrix} \end{bmatrix} \boldsymbol{\xi}(t), \tag{45}
\end{aligned}$$

$$\begin{aligned}
\mathbf{0} & = 2 [\mathbf{e}^T(t) + \dot{\mathbf{e}}^T(t) + \mathbf{e}^T(t_k)] \mathbf{G}^T \times \\
& \quad [-\dot{\mathbf{e}}(t) + \mathbf{A}\mathbf{e}(t) + \mathbf{B}\mathbf{e}(t - \tau(t)) + \mathbf{E}\mathbf{g}(\mathbf{D}\mathbf{e}(t)) - \mathbf{K}\mathbf{C}\mathbf{e}(t - h(t)) + \mathbf{K}\mathbf{C}\boldsymbol{\varepsilon}(t_s)] = \\
& \quad 2 [\mathbf{e}^T(t) + \dot{\mathbf{e}}^T(t) + \mathbf{e}^T(t - h(t)) - \boldsymbol{\varepsilon}^T(t_s)] \mathbf{G}^T \times
\end{aligned}$$

$$[-\dot{e}(t) + Ae(t) + Be(t - \tau(t)) + Eg(De(t)) - KCe(t - h(t)) + KC\varepsilon(t_s)], \tag{46}$$

$$\mathbf{0} \leq -[g(De(t)) - W^+ De(t)]^T F[g(De(t)) - W^- De(t)], \tag{47}$$

$$\dot{V}(t) \leq \xi^T(t)[\Psi_1 + \Psi_2 + \Psi_3(\tau(t)) + \Psi_4(h(t)) + \Psi_5(\tau(t)) + \Psi_6(h(t))]\xi(t). \tag{48}$$

基于凸组合技术, $\Psi_1 + \Psi_2 + \Psi_3(\tau(t)) + \Psi_4(h(t)) + \Psi_5(\tau(t)) + \Psi_6(h(t)) \leq \mathbf{0}$ 成立, 当且仅当以下四式成立:

$$\Psi_1 + \Psi_2 + \Psi_3(\tau_1) + \Psi_4(d) + \Psi_5(\tau_1) + \Psi_6(d) < \mathbf{0}, \tag{49}$$

$$\Psi_1 + \Psi_2 + \Psi_3(\tau_1) + \Psi_4(h) + \Psi_5(\tau_1) + \Psi_6(h) < \mathbf{0}, \tag{50}$$

$$\Psi_1 + \Psi_2 + \Psi_3(\tau_2) + \Psi_4(d) + \Psi_5(\tau_2) + \Psi_6(d) < \mathbf{0}, \tag{51}$$

$$\Psi_1 + \Psi_2 + \Psi_3(\tau_2) + \Psi_4(h) + \Psi_5(\tau_2) + \Psi_6(h) < \mathbf{0}. \tag{52}$$

则有 $\dot{V}(t) < 0$, 根据 Schur 补引理, $\dot{V}(t) < 0$, 等价于线性矩阵不等式(14)~(17), 所以误差系统(6)是渐近稳定的, 证明结束.

3 数值仿真

为了验证本文所提方法的有效性, 做如下的仿真:

$$\begin{cases} \dot{x}_1(t) = a(x_2(t) - m_1x_1(t) + g(x_1(t))) - cx_1(t - d(t)), \\ \dot{x}_2(t) = x_1(t) - x_2(t) + x_3(t) - cx_1(t - d(t)), \\ \dot{x}_3(t) = -bx_2(t) + c(2x_1(t - d(t)) - x_3(t - d(t))), \end{cases}$$

其中, 非线性函数 $g(x_1(t)) = (1/2)(m_1 - m_0)(|x_1(t) + 1| - |x_1(t) - 1|)$, 其余参数 $m_0 = -1/7, m_1 = 2/7, a = 9, b = 14.28, c = 0.1$, 时变时滞 $d(t) = |\sin(t)|$.

系统(1)和(2)可表示成 Lur'e 系统的形式:

$$\mathbf{A} = \begin{bmatrix} -am_1 & a & 0 \\ 1 & -1 & 1 \\ 0 & -b & 0 \end{bmatrix}, \mathbf{B} = \begin{bmatrix} -c & 0 & 0 \\ -c & 0 & 0 \\ 2c & 0 & -c \end{bmatrix}, \mathbf{E} = \begin{bmatrix} -a(m_0 - m_1) \\ 0 \\ 0 \end{bmatrix}, \mathbf{C} = \mathbf{D} = \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}^T,$$

非线性函数 $f_1(x_1(t)) = \frac{1}{2}(|x_1(t) + 1| - |x_1(t) - 1|) \in [0, 1], f_2(x_2(t)) = f_3(x_3(t)) = 0$.

主从系统的初始值分别取为 $\mathbf{x}(0) = [0.2 \ 0.3 \ 0.2]^T, \mathbf{y}(0) = [-0.3 \ -0.1 \ 0.4]^T$. 取 $\mu_1 = \mu_2 = 5, \kappa_1 = \kappa_2 = 2, u_1 = u_2 = 0, \tau_2 = 1, \sigma_1(0) = 0.454, \sigma_2(0) = 0.456$, 应用 MATLAB 求解线性矩阵不等式(14)~(17), 根据定理 1 计算所得到的最大采样区间的数值如表 1 所示. 由表 1 可知, 与文献[1, 4, 13, 16]相比, 本文所得到的最大采样区间的数值更优. 当 $h = 0.5629$ 时, 求解线性矩阵不等式(14)~(17), 所得控制器增益矩阵和触发参数如下:

$$\mathbf{W}_1 = \begin{bmatrix} 0.8592 & 0.1502 & -0.2593 \\ 0.1502 & 0.8432 & -0.1215 \\ -0.2593 & -0.1215 & 1.1246 \end{bmatrix}, \mathbf{W}_2 = \begin{bmatrix} 0.7442 & -0.2495 & 0.3496 \\ -0.2495 & 0.9036 & 0.2496 \\ 0.3496 & 0.2496 & 0.9733 \end{bmatrix},$$

$$\mathbf{W}_3 = \begin{bmatrix} 0.7245 & -0.2505 & 0.3512 \\ -0.2505 & 0.8986 & 0.2484 \\ 0.3512 & 0.2484 & 0.9816 \end{bmatrix}, \mathbf{K} = \begin{bmatrix} 3.3827 \\ 0.3015 \\ -2.9116 \end{bmatrix}.$$

表 1 最大采样区间

Table 1 The maximum sampling period h

reference	[1]	[4]	[13]	[16]	theorem
h	0.4824	0.4355	0.4789	0.4371	0.5629

MATLAB 仿真得到的误差向量和传输间隔的响应曲线, 如图 1 和图 2 所示. 从图 1 可以看出, 在最大采样区间 $h = 0.5629$ 时, 误差系统的状态轨迹在有限时间内收敛于零, 即时滞 Lur'e 系统在短时间内能够实现主从系统同步. 从图 2 可以看出, 在最大采样区间 $h = 0.5629$ 时, 主从同步过程中加入事件触发机制能够减

少信号的传输次数,提高通信利用率。

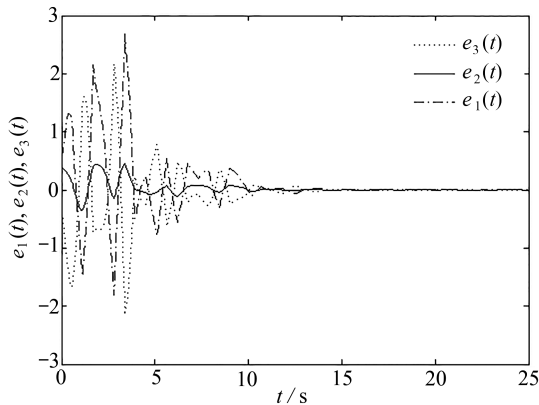


图1 采样周期为0.562 9时,误差 $e(t)$ 的响应曲线

Fig. 1 Responses of $e(t)$ with a sampling period of $h = 0.5629$

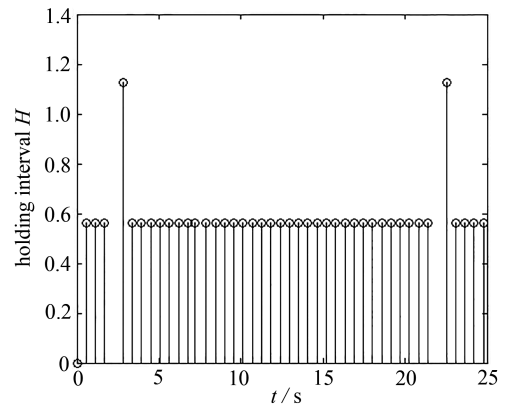


图2 事件触发传输瞬间及其释放间隔

Fig. 2 Transmission instants and releasing intervals in the case of event triggering

4 结 论

本文研究了基于事件触发机制的混沌Lur'e系统主从同步问题,为了减少保守性,在构造Lyapunov-Krasovskii泛函时考虑了非线性函数的扇区约束条件,引进了基于辅助函数的积分不等式和增广Lyapunov-Krasovskii泛函,结合Lyapunov稳定性理论和线性矩阵不等式技术,给出了混沌Lur'e系统主从同步的充分条件.通过设置触发条件能够降低系统的状态信息在通信网络中的传输次数,缓解网络带宽压力.最后,通过数值仿真验证了所提方法的有效性。

致谢 本文作者衷心感谢上海工程技术大学研究生科研创新项目(0232-E3-0903-19-01286)对本文的资助。

参考文献(References):

- [1] GE C, WANG B F, PARK J H, et al. Improved synchronization criteria of Lur'e systems under sampled-data control[J]. *Nonlinear Dynamics*, 2018, **94**(4): 2827-2839.
- [2] ZHANG R M, ZENG D Q, LIU X Z, et al. A new method for quantized sampled-data synchronization of delayed chaotic Lur'e systems[J]. *Applied Mathematical Modelling*, 2019, **70**: 471-489.
- [3] LEE T H, PARK J H. Improved criteria for sampled-data synchronization of chaotic Lur'e systems using two new approaches[J]. *Nonlinear Analysis: Hybrid Systems*, 2017, **24**: 132-145.
- [4] HUA C, GE C, GUAN X P. Synchronization of chaotic Lur'e systems with time delays using sampled-data control[J]. *IEEE Transactions on Neural Networks and Learning Systems*, 2015, **26**(6): 1214-1221.
- [5] JI X F, ZU C, SU H Y. Delay-dependent synchronisation for singular Lur'e systems using time delay feedback control[J]. *International Journal of Modelling Identification & Control*, 2013, **19**(2): 125-133.
- [6] LI T, SONG A, FEI S. Master-slave synchronization for delayed Lur'e systems using time-delay feedback control[J]. *Asian Journal of Control*, 2015, **13**(6): 879-892.
- [7] ZHONG M Y, HAN Q L. Fault-tolerant master-slave synchronization for Lur'e systems using time-delay feedback control[J]. *IEEE Transactions on Circuits and Systems I: Regular Papers*, 2009, **56**(7): 1391-1404.
- [8] FAN, Y Q, WANG W, LIU Y. Synchronization for a class of chaotic systems based on adaptive control design of input-to-state stability[J]. *International Journal of Innovative Computing, Information & Control*, 2015, **11**(3): 803-814.
- [9] LU J, CAO J, HO D W C. Adaptive stabilization and synchronization for chaotic Lur'e systems with time-varying delay[J]. *IEEE Transactions on Circuits & Systems I: Regular Papers*, 2008, **55**(5): 1347-1356.
- [10] CHEN W H, DAN W, WANG Z P, et al. Master-slave synchronization of chaotic Lur'e systems under delayed

- impulsive control[C]//*Proceedings of the 31st Chinese Control Conference*. Hefei, 2012: 1345-1350.
- [11] YIN C, ZHONG S M, CHEN W F. Design PD controller for master-slave synchronization of chaotic Lur'e systems with sector and slope restricted nonlinearities[J]. *Communications in Nonlinear Science & Numerical Simulation*, 2011, **16**(3): 1632-1639.
- [12] 严欢, 高岩波. 混沌 Lur'e 系统基于时滞反馈 PD 控制的同步[J]. 南通大学学报(自然科学版), 2017, **16**(4): 12-21.(YAN Huan, GAO Yanbo. Synchronization for chaotic Lur'e systems via delayed feedback PD control [J]. *Journal of Nantong University(Natural Science Edition)*, 2017, **16**(4): 12-21.(in Chinese))
- [13] ZHANG R M, ZENG D, ZHONG S M. Novel master-slave synchronization criteria of chaotic Lur'e systems with time delays using sampled-data control[J]. *Journal of the Franklin Institute*, 2017, **354**(12): 4930-4954.
- [14] DUAN W, LI Y, FU X R. Improved synchronization criteria for time-delayed chaotic Lur'e systems using sampled-data control[J]. *Journal of Physics: Conference Series*, 2017, **814**: 12-22.
- [15] JUNMIN P, POOGYEON P. H_∞ sampled-state feedback control for synchronization of chaotic Lur'e systems with time delays[J]. *Journal of the Franklin Institute*, 2018, **355**(16): 8005-8026.
- [16] SHANGGUAN X C, HE Y, LIN W J, et al. Improved synchronization of chaotic Lur'e systems with time delay using sampled-data control[J]. *Journal of the Franklin Institute*, 2017, **354**(3): 1618-1636.
- [17] LIU S J, ZHOU L. Network synchronization and application of chaotic Lur'e systems based on event-triggered mechanism[J]. *Nonlinear Dynamics*, 2016, **83**(4): 2497-2507.
- [18] 刘晨, 刘磊. 基于事件触发策略的多智能体系统的最优主-从一致性分析[J]. 应用数学和力学, 2019, **40**(11): 1278-1288.(LIU Chen, LIU Lei. Optimal leader-follower consensus of multi-agent systems based on the event-triggered strategy[J]. *Applied Mathematics and Mechanics*, 2019, **40**(11): 1278-1288.(in Chinese))
- [19] HEEMELS W P M H, JOHANSSON K H, TABUADA P. An introduction to event-triggered and self-triggered control[C]//*51st IEEE Conference on Decision and Control (CDC)*. Maui, HI, 2012: 3270-3285.
- [20] GU Z, SHI P, YUE D, et al. Decentralized adaptive event-triggered H_∞ filtering for a class of networked nonlinear interconnected systems[J]. *IEEE Transactions on Cybernetics*, 2019, **49**(5): 1570-1579.
- [21] WANG Y Y, KARIMI H R, YAN H C. An adaptive event-triggered synchronization approach for chaotic Lur'e systems subject to aperiodic sampled data[J]. *IEEE Transactions on Circuits and Systems II: Express Briefs*, 2019, **66**(3): 442-446.
- [22] PARK P G, LEE W I, LEE S Y. Auxiliary function-based integral inequalities for quadratic functions and their applications to time-delay systems[J]. *Journal of the Franklin Institute*, 2015, **352**(4): 1378-1396.
- [23] SEURET A, GOUAISBAUT F. Wirtinger-based integral inequality: application to time-delay systems[J]. *Automatica*, 2013, **49**(9): 2860-2866.
- [24] LIU K, SEURET A, XIA Y Q. Stability analysis of systems with time-varying delays via the second-order Bessel-Legendre inequality[J]. *Automatica*, 2017, **76**(2): 138-142.
- [25] ZHANG C K, HE Y, JIANG L, et al. An extended reciprocally convex matrix inequality for stability analysis of systems with time-varying delay[J]. *Automatica*, 2017, **85**: 481-485.